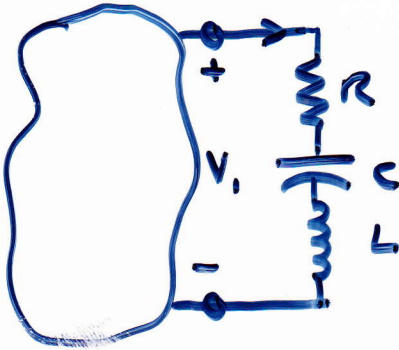


Resonant Circuits.

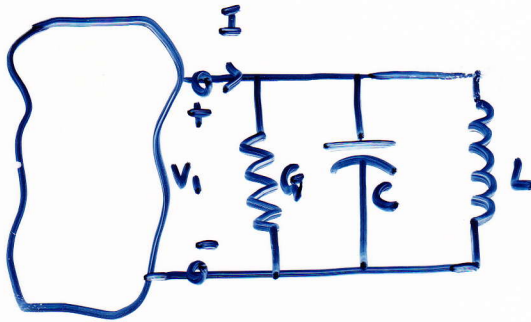
Consider RLC circuits

SERIES



$$\bullet \underline{Z}(j\omega) = R + j\omega L + \frac{1}{j\omega C} \quad (1)$$

PARALLEL



$$\bullet \underline{Y}(j\omega) = G + j\omega C + \frac{1}{j\omega L} \quad (2)$$

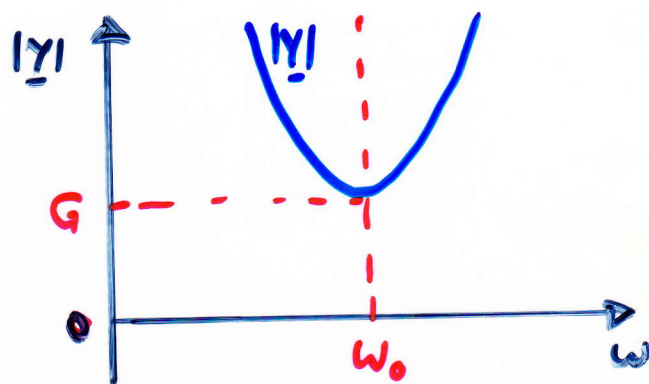
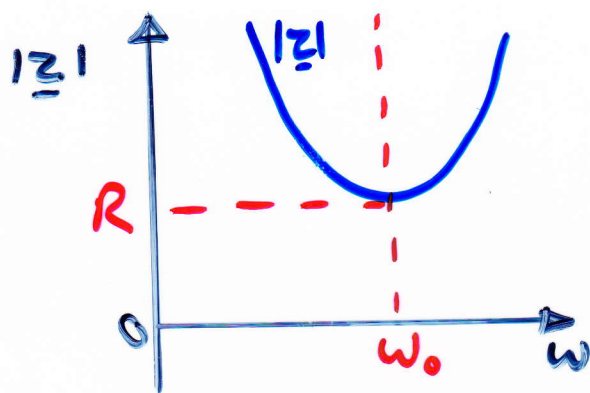
Equations (1) and (2) have same basic form

If $\omega L = \frac{1}{\omega C}$ then imaginary terms equal zero.

$$\therefore \omega_0 = \frac{1}{\sqrt{LC}}$$

RESONANT FREQUENCY

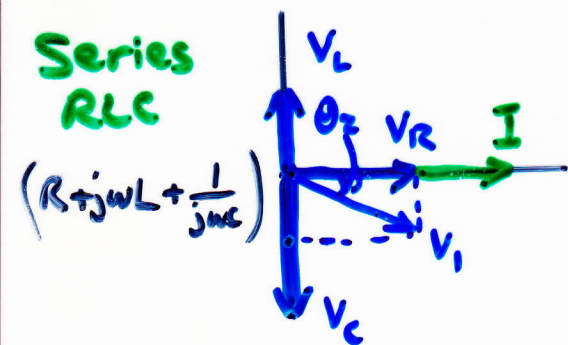
At this freq. $\underline{Z}(j\omega_0) = R$ for SERIES circuits
 $\underline{Y}(j\omega_0) = G$ " PARALLEL "



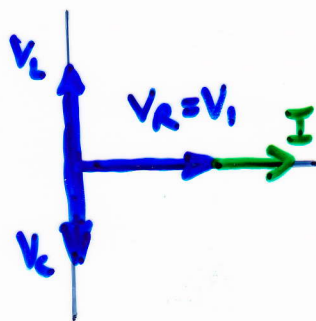
At resonance voltage and current in phase, therefore phase angle is zero.

$\omega < \omega_0$

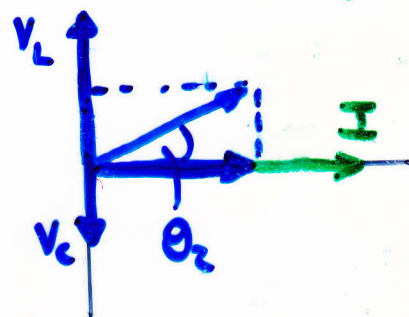
Series
RLC



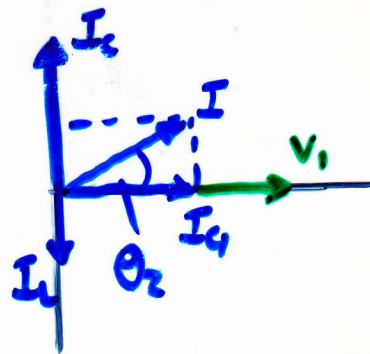
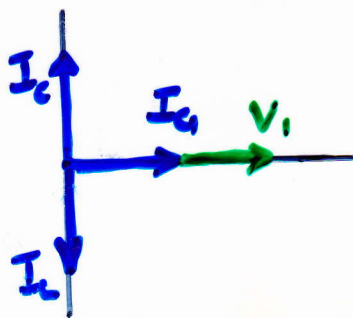
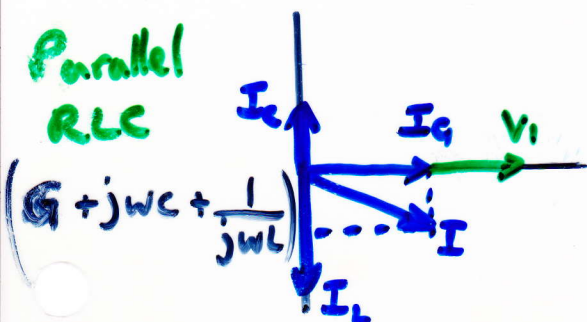
$\omega = \omega_0$



$\omega > \omega_0$



Parallel
RLC



Q

Consider series RLC. Can define a quantity known as the quality factor, Q :

$$Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 C R} = \frac{1}{R} \sqrt{\frac{L}{C}} \quad (3)$$

[Question: what would be the similar expression for parallel circuit?]

Considering (3), if R is small the Q is high, if R is large Q is low.

To look at Q graphical let us do a little more mathematical manipulation.

Consider the admittance of the series circuit.

$$\begin{aligned} \underline{Y}(j\omega) &= \frac{1}{R(1 + j(\frac{1}{R})(\omega L - 1/\omega C))} \\ &= \frac{1}{R(1 + j\frac{\omega L}{R} - j/\omega C R)} \end{aligned}$$

$$= \frac{1}{R \left[1 + jQ \left(\frac{\omega L}{R} - \frac{1}{\omega C R} \right) \right]}$$

Since $Q = \omega_0 L / R = 1 / \omega_0 C R$

Then $\underline{Y}(j\omega) = \frac{1}{R \left(1 + jQ \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right) \right)}$

Since $\underline{I} = \underline{Y} \underline{V}_i$ & $\underline{V}_R = \underline{I} R$

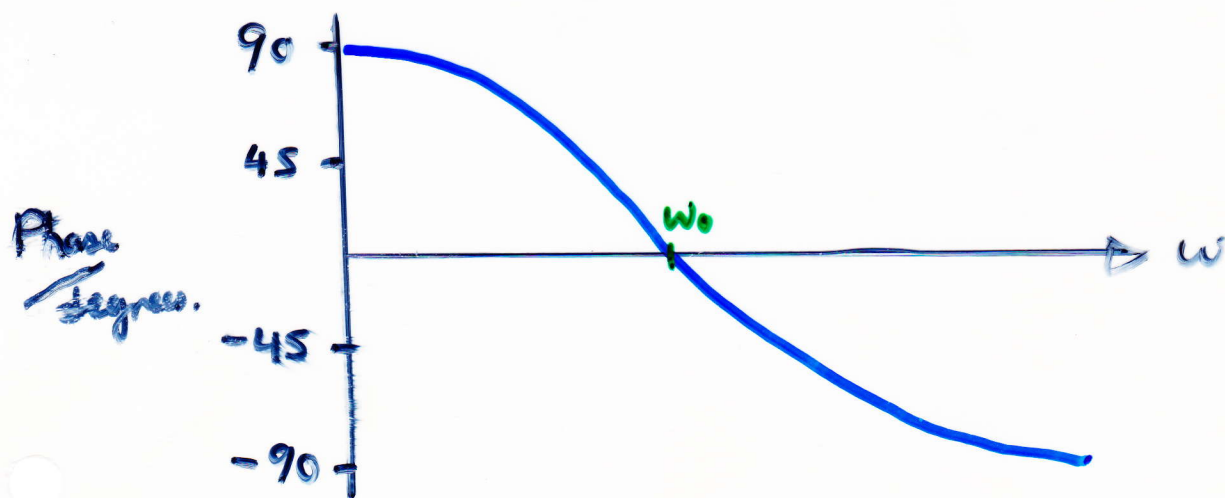
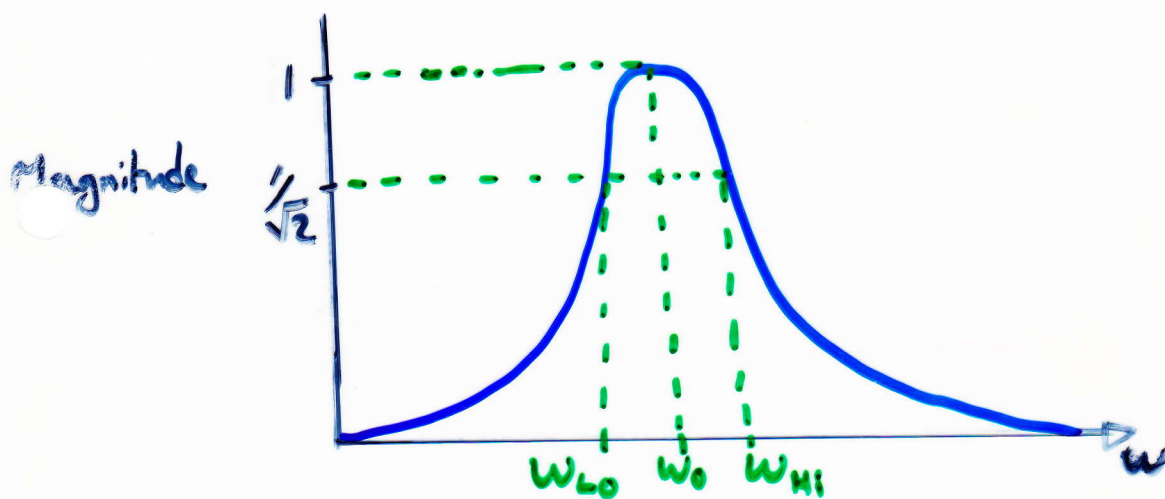
then $\frac{\underline{V}_R}{\underline{V}_i} = \underline{G}_V(j\omega) = \frac{1}{1 + jQ \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)}$

↑
Transfer function.

Looking at magnitude & phase

$$M(\omega) = \frac{1}{[1 + Q^2(\omega/\omega_0 - \omega_0/\omega)^2]^{\frac{1}{2}}}$$

$$\phi(\omega) = -\tan^{-1} Q \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)$$



$$\text{Let magnitude} = \frac{1}{\sqrt{2}} = \left| \frac{1}{1 + jQ(\omega/\omega_0 - \omega_0/\omega)} \right|$$

$$\therefore Q \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right) = \pm 1$$

Solving this can see have four freq.

$$\omega = \pm \frac{\omega_0}{2Q} \pm \omega_0 \sqrt{\left(\frac{1}{2Q} \right)^2 + 1}$$

Taking +ve freq.

$$\therefore \omega_{Lo} = \omega_0 \left[-\frac{1}{2Q} + \sqrt{\left(\frac{1}{2Q} \right)^2 + 1} \right] \quad (4)$$

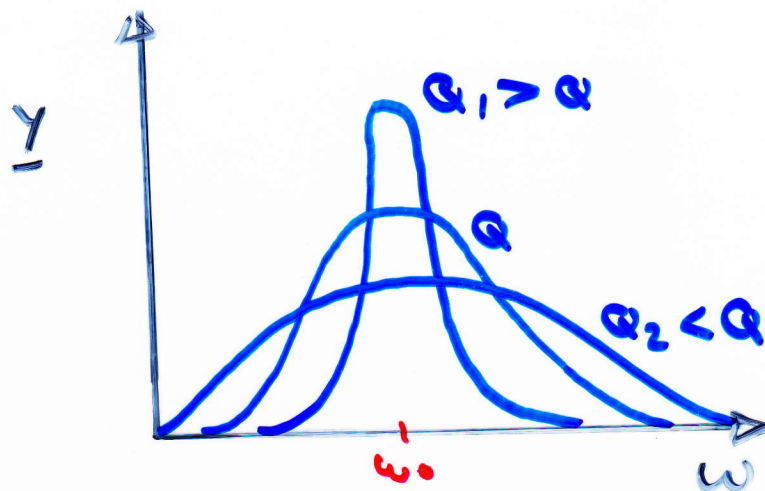
$$\omega_{Hi} = \omega_0 \left[\frac{1}{2Q} + \sqrt{\left(\frac{1}{2Q} \right)^2 + 1} \right] \quad (5)$$

Subtracting (4) from (5)

$$\text{Bandwidth, } BW = \omega_{Hi} - \omega_{Lo} = \frac{\omega_0}{Q}$$

Multiplying (4) & (5)

$$\omega_0^2 = \omega_{Lo} \omega_{Hi}$$



Q also defined as

$$Q = 2\pi \frac{W_s}{W_D}$$

GENERAL DEFN

W_D energy dissipated per cycle

W_s max. energy stored at resonance

7th

9th

Example.

Irwin.

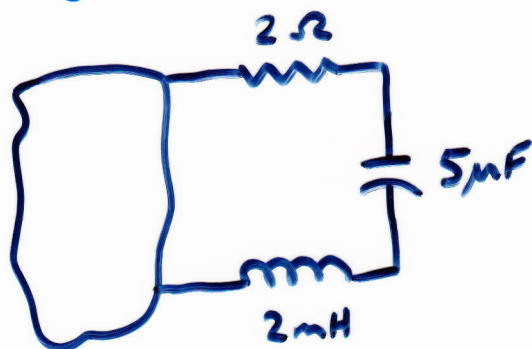
Ex 11.9

12.9

Series circuit with $R = 2\Omega$, $L = 2\text{mH}$
and $C = 5\mu\text{F}$.

Determine the resonant freq., the quality factor and bandwidth.

What is the change in Q and BW if R is changed from 2Ω to 0.2Ω ?



$$\omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{(2 \times 10^{-3} \times 5 \times 10^{-6})^{1/2}}$$

$$= 10^4 \text{ rad/s}$$

$$\therefore f_0 = \frac{10^4}{2\pi} = \underline{1592 \text{ Hz.}}$$

$$Q = \frac{\omega_0 L}{R} = \frac{10^4 \times 2 \times 10^{-3}}{2} = \underline{10}$$

$$BW = \frac{\omega_0}{Q} = \frac{10^4}{10} = \underline{10^3} \text{ rad/s.}$$

If R becomes 0.2Ω then:

$$Q = \frac{10^4 \times 2 \times 10^{-3}}{0.2 \times 0.1}$$

$$= 100$$

$$BW = \frac{\omega_0}{Q} = \frac{10^4}{10^2} \\ = \underline{10^2} \text{ rad/s.}$$